

# IBIS-AMI Modeling and Simulation of DMT in Preparation for 448Gbps Applications

Fangyi Rao, Keysight Technologies

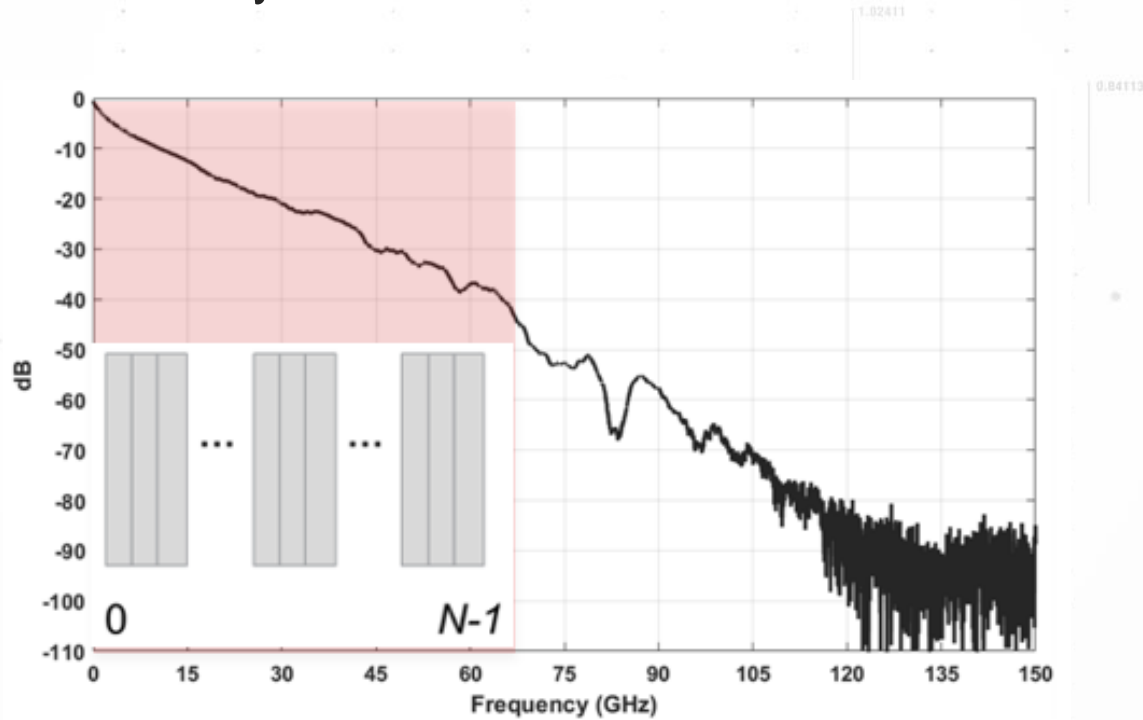
Hongtao Zhang, Jinbiao Xu, Geoff Zhang, AMD



DesignCon Hybrid IBIS Summit  
Santa Clara, CA  
January 31, 2025

# Introduction

- Unlike PAM signaling, DMT (discrete multi-tone) divides the baseband channel into multiple sub-channels
- Now, each sub-channel can carry different number of bits based on the robustness of the sub-channel



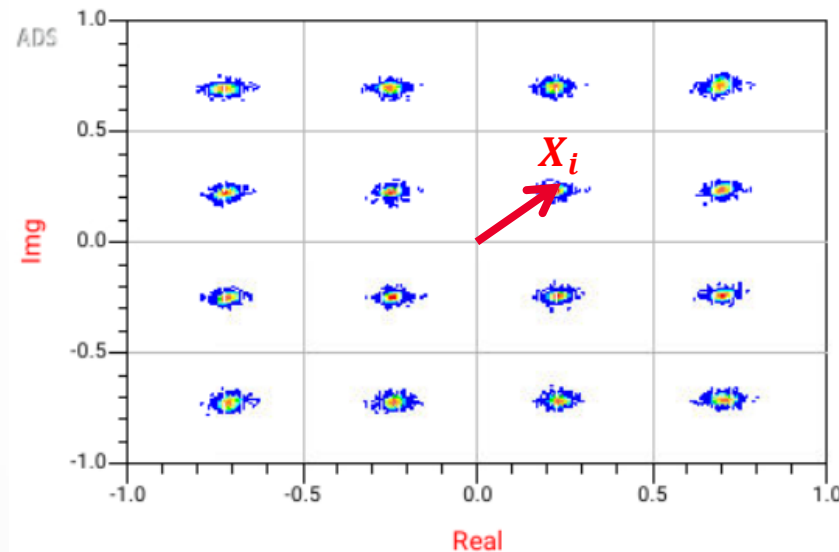
- The focus of this presentation is to introduce IBIS-AMI modeling for link channels using DMT

# Review of QAM (Quadrature Amplitude Modulation)

- QAM-N signal ( $N = 2^M$ )

$$S(t) = \text{Re}[X_i \cdot e^{j2\pi f_c t}] = \text{Re}[X_i] \cdot \cos(2\pi f_c t) - \text{Im}[X_i] \cdot \sin(2\pi f_c t) = I(t) + Q(t), \quad i = 1, \dots, N$$

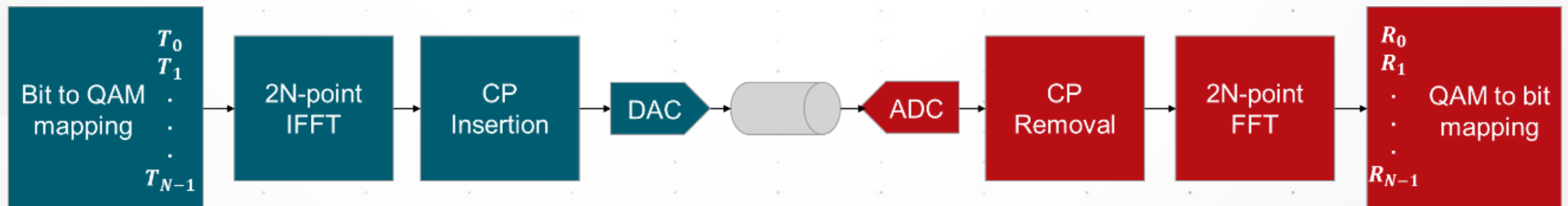
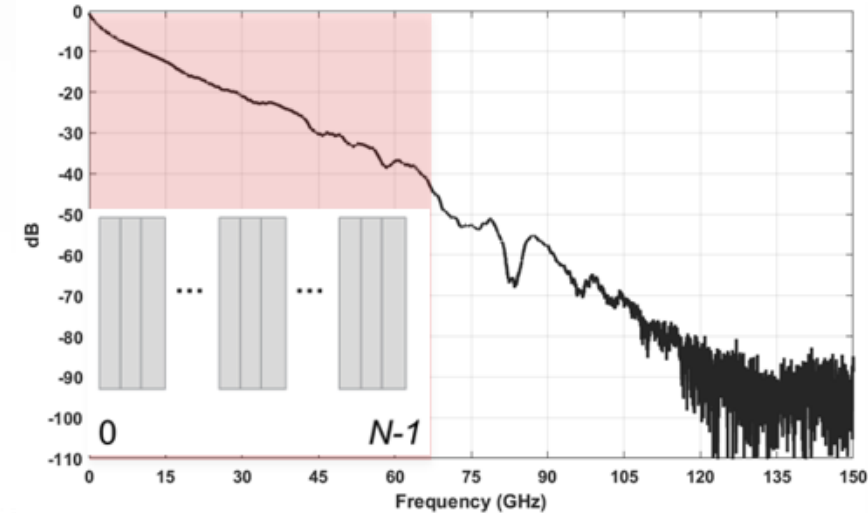
- TX maps each M-bit sequence to a unique complex amplitude  $X_i$  (QAM symbol) and modulates  $I$  and  $Q$  signals with  $X_i$
- RX recovers  $X_i$  from  $I$  and  $Q$  amplitudes and maps  $X_i$  to the M-bit sequence
- Each QAM symbol carries M bits



QAM16 example: each QAM16 symbol represents a 4-bit sequence

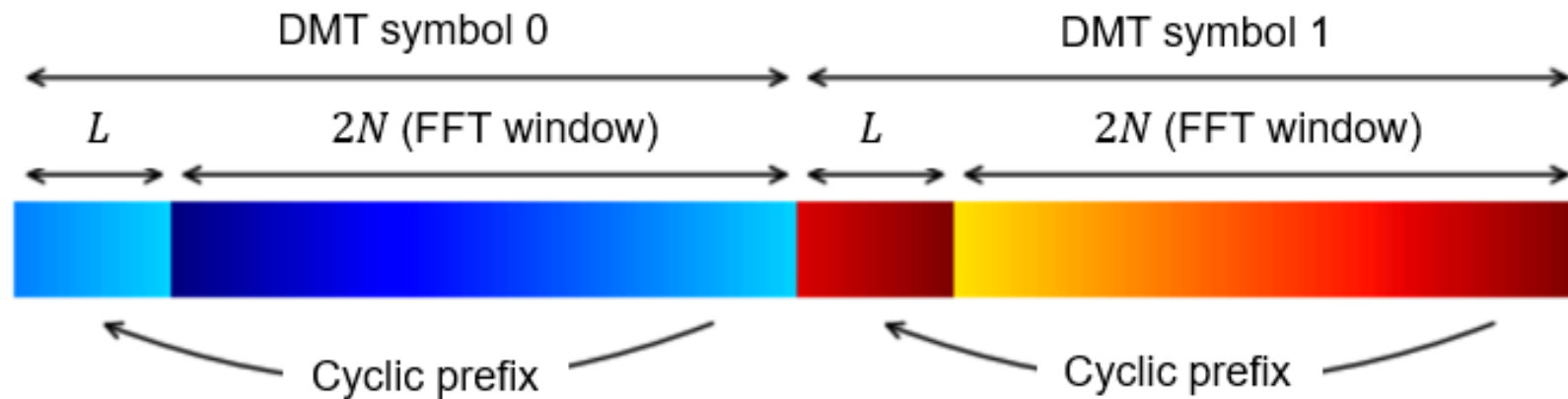
# Key Components in DMT

1. Divide the channel into  $N$  sub-channels, from DC to  $f_{\max}$
2. Assign modulation order to each sub-channel,  $QAM(k)$ 
  - For a given time frame, generate the complex values,  $T_0, T_1, \dots, T_{N-1}$
3. Extend to the negative frequency to form  $2N$ -point complex array
4. Do  $2N$ -point IFFT to produce a valued array:  $d_0, d_1, \dots, d_{2N-1}$
5. Insert CP of length  $L$ :  $d_{2N-L}, \dots, d_{2N-1}, d_0, d_1, \dots, d_{2N-1}$
6. Transmit the  $2N + L$  samples
7. Remove the first  $L$  symbols from the received symbols
8. Apply  $2N$ -point FFT the remaining to obtain  $R_0, R_1, \dots, R_{2N-1}$
9. Detect and map  $R_k$  back to binary data



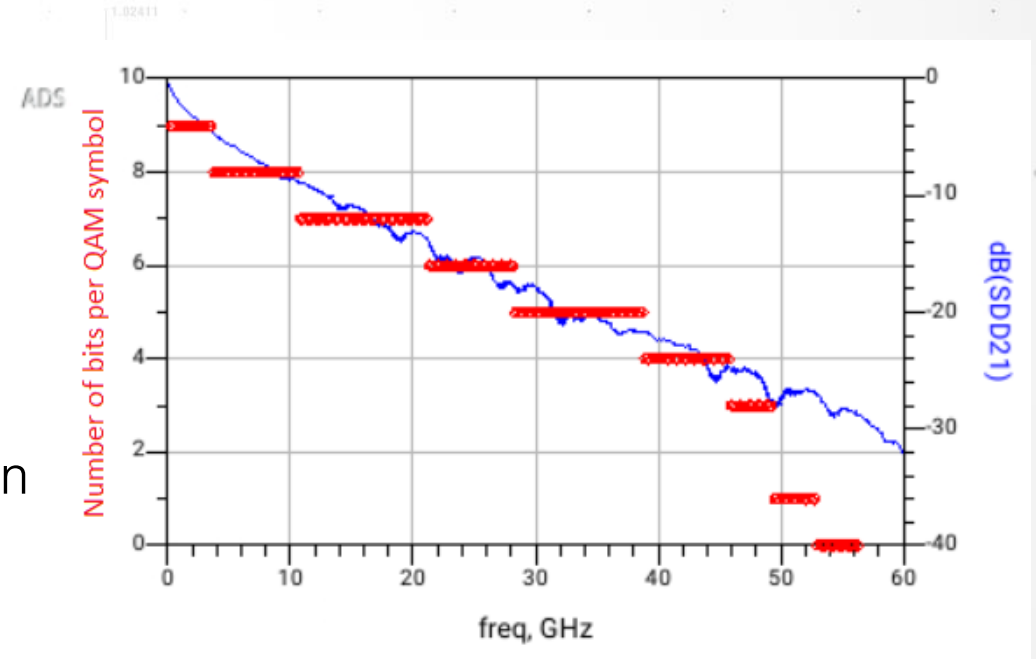
# Cyclic Prefix (CP) and Why

- Recall that
  - Multiplication in the frequency domain corresponds to circular convolution in the time domain; likewise, circular convolution in the time domain corresponds to multiplication in the frequency domain
  - The channel impact on the transmitted signal can be presented as a linear convolution of the channel impulse response with the transmitted signal
- Therefore, we would need some way of turning the channel response into a circular convolution, not a linear convolution. This is how we can do it:
  - For each time-domain block generated by the transmitter, a CP is inserted before it. The CP takes the last  $L$  samples of the  $2N$ -sample time-domain symbol and concatenates it to the front of the block to be transmitted



# DMT Advantages

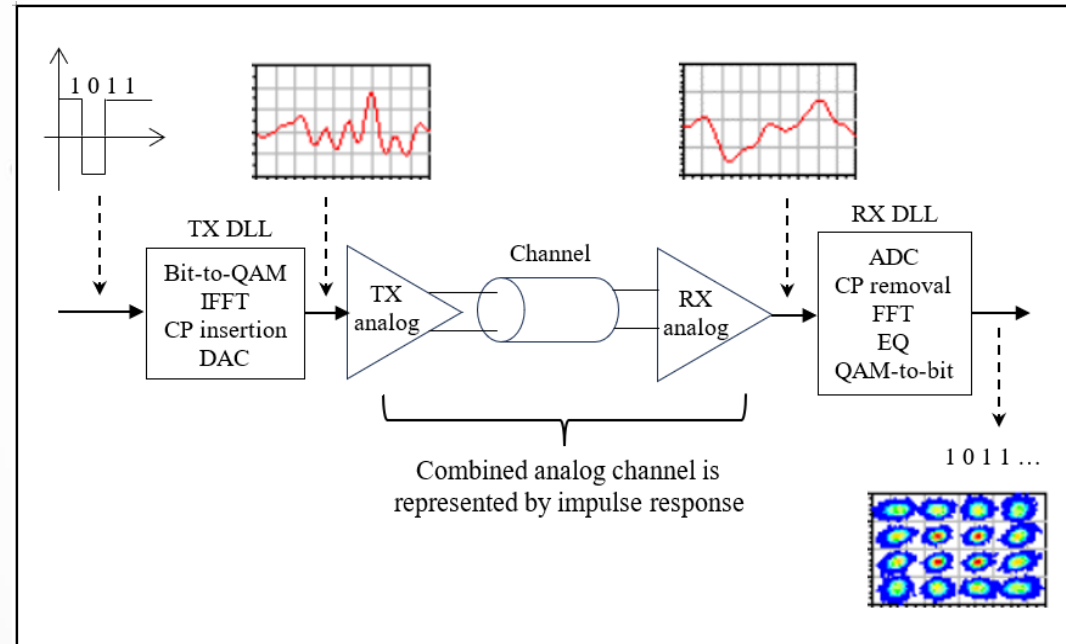
- Bit loading
  - Assign different numbers of bits to different sub-channels
  - The higher the sub-channel SNR, the more bits are assigned to it
  - Maximized bandwidth efficiency
  - Tolerant of channel spectral notch/discontinuity
- RX equalization
  - CP renders the channel effect a circular convolution
  - Each sub-channel is a narrow band signal
  - RX equalization is as simple as a scalar multiplication in frequency domain



# AMI Modeling for DMT Proposal (1)

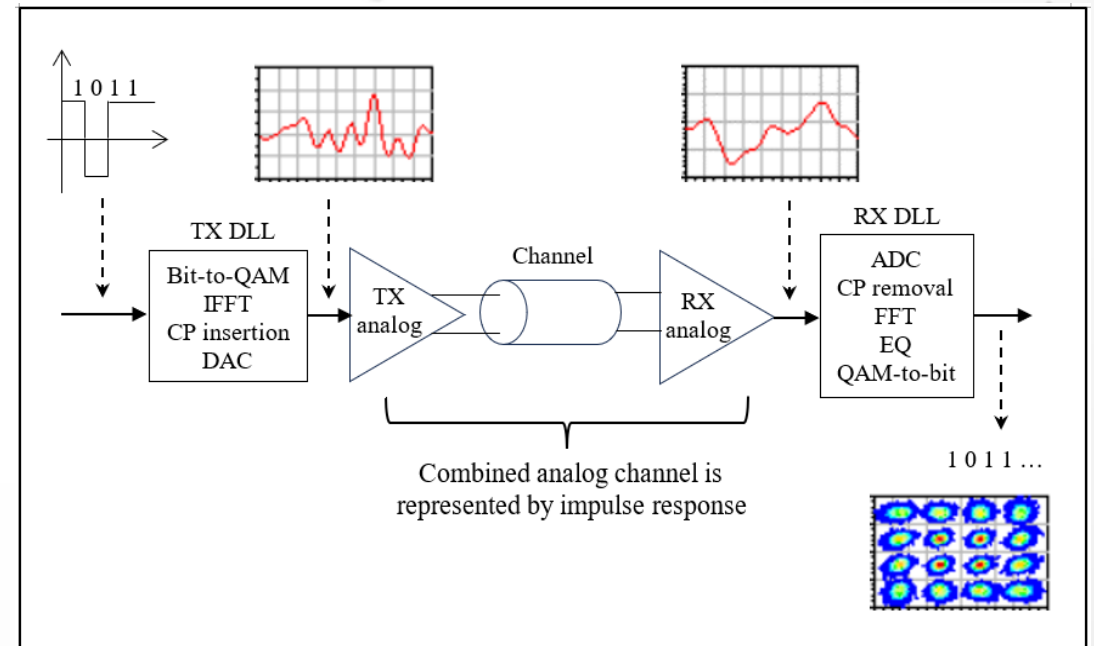
- TX DLL encapsulates
  - sample rate and sub-channel number determinations
  - bit loading
  - bit-to-QAM mapping
  - IFFT
  - CP insertion
  - DAC
  - TX front-end

- RX DLL encapsulates
  - timing synchronization
  - ADC
  - CP removal
  - FFT
  - Equalization
  - QAM-to-bit mapping
  - bit data recovery



# AMI Modeling for DMT Proposal (2)

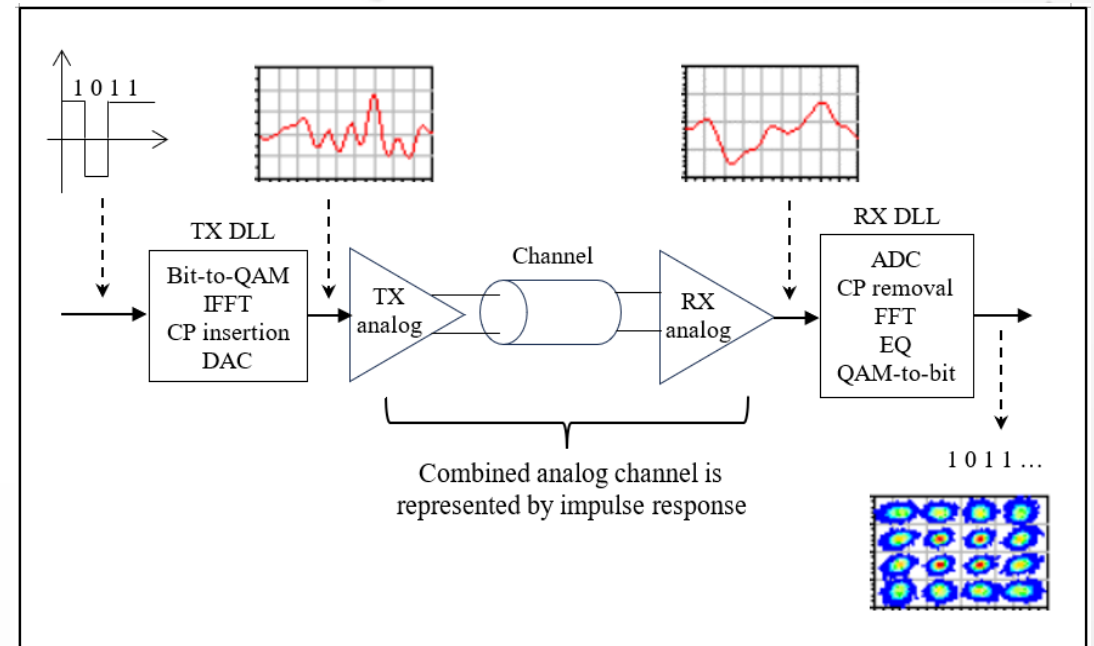
- TX AMI\_Init
  - Use the inverse of the target data rate for the symbol\_time argument
  - Based on the data rate and the input impulse response of the analog channel, TX AMI\_Init determines configurations such as sample rate, sub-channel number, bit loading, CP length, etc.
  - TX AMI\_Init writes the configuration data to a file to be read by the RX AMI\_Init
  - To distinguish files used by different TX-RX pairs, the file name can be passed into TX and RX DLLs by a Model Specific parameter of Usage "In" and Type "String"
  - EDA tool or model user is responsible for assigning the same file name to pairing TX and RX models





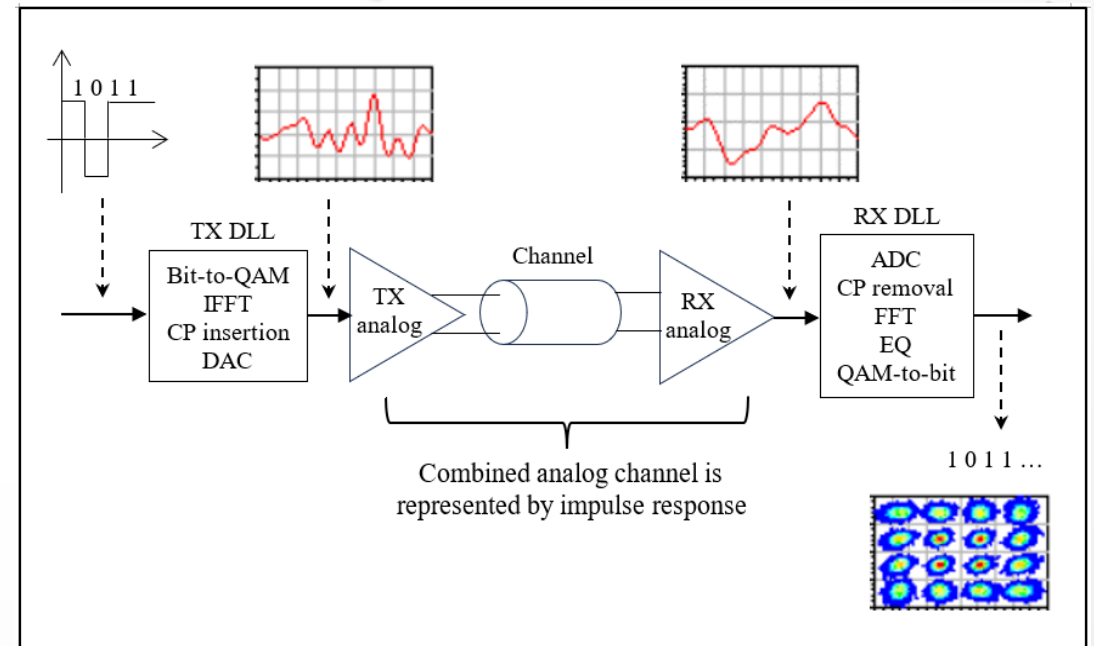
# AMI Modeling for DMT Proposal (3)

- RX AMI\_Init
  - RX AMI\_Init reads the file and retrieves the TX configuration
  - Model developer is responsible for model interoperability when TX and RX exchange DMT configuration data through the file
  - Model developer is also responsible for the consistency between TX bit-to-QAM mapping and RX QAM-to-bit mapping
- As in TX AMI\_Init, use the inverse of the target data rate for the symbol\_time argument in RX AMI\_Init



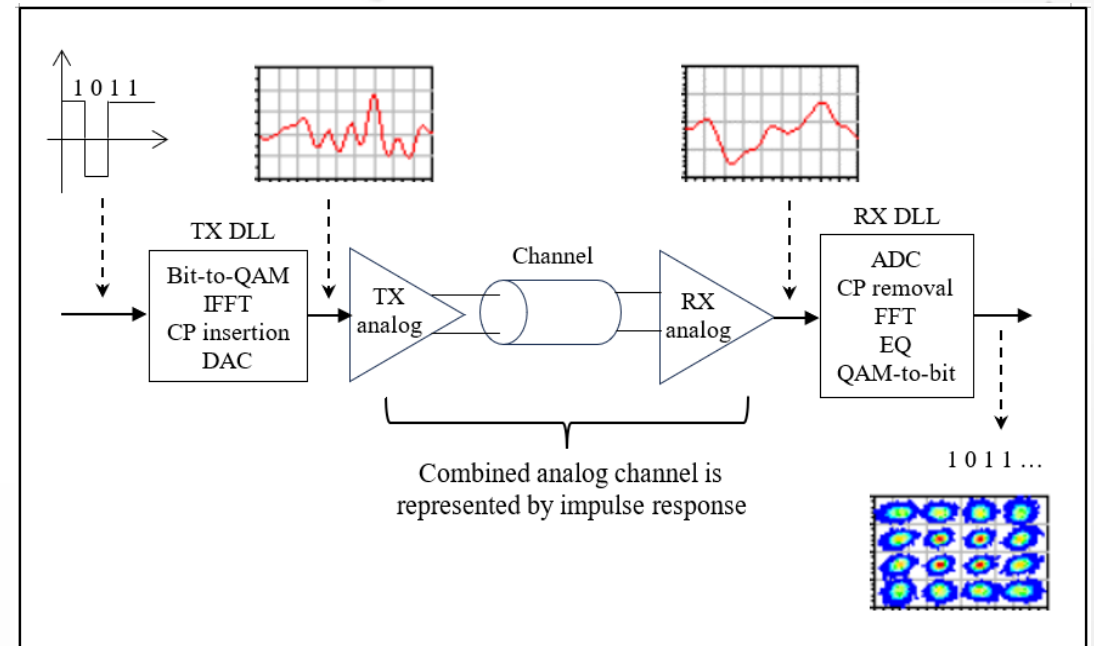
# AMI Modeling for DMT Proposal (4)

- Another possible mechanism for exchanging configuration information between TX and RX
  - TX AMI\_Init writes the configuration data to a new Reserved parameter and returns it in the AMI\_parameters\_out output string
  - EDA tool extracts the parameter value and includes it in the AMI\_parameters\_in input string of RX AMI\_Init
  - The parameter is of Usage “InOut” and Type “String”
  - The string content and format of this parameter is proprietary between TX and RX models
- Both approaches provide the same flexibility for model development (the file approach is used in this paper)



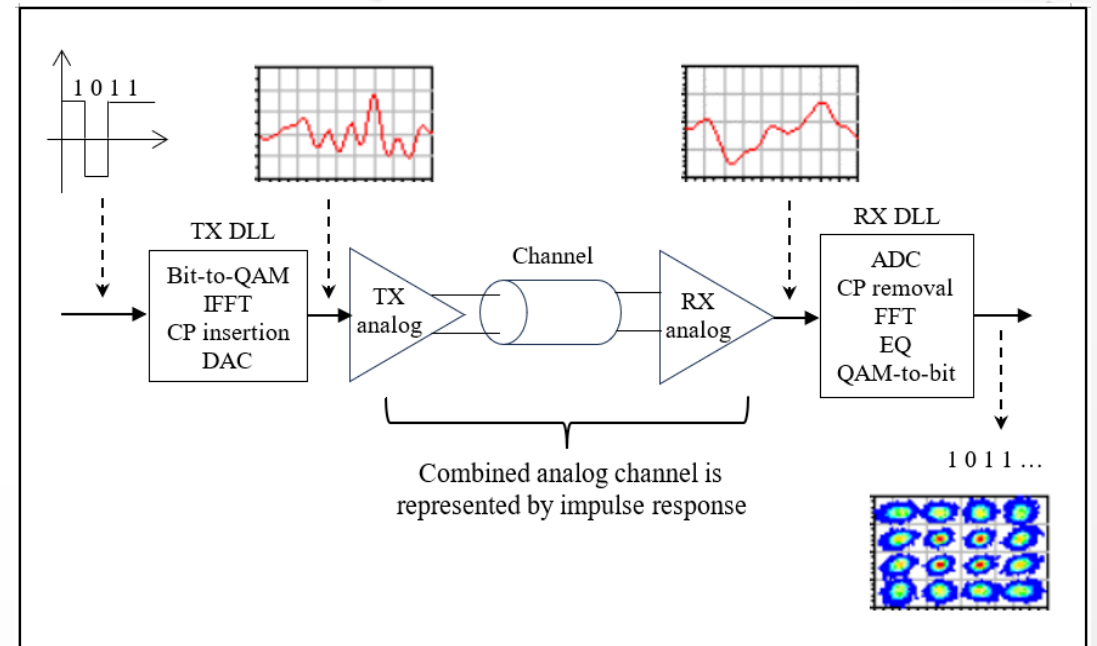
# AMI Modeling for DMT Proposal (5)

- TX AMI\_GetWave
  - The input waveform is a binary waveform at the target data rate generated by the simulator
  - TX AMI\_GetWave performs bit-to-QAM conversion, IFFT, CP insertion and DAC and returns the DMT symbol (including CP) waveform
  - TX jitters, if present, need to be injected by TX AMI\_GetWave, and the input waveform should be free of jitter
- The AMI\_GetWave block size is not required to be a multiple of the DMT symbol (including CP) length



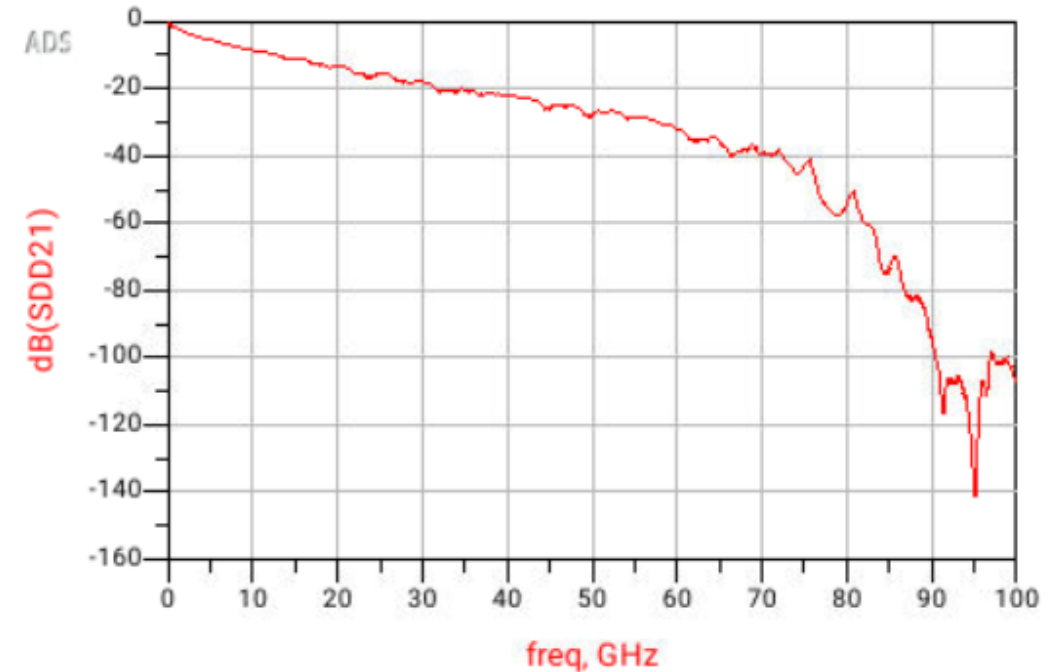
# AMI Modeling for DMT Proposal (6)

- RX AMI\_GetWave
  - performs timing synchronization, ADC, CP removal, FFT, equalization, QAM-to-bit mapping, and bit data recovery
  - returns recovered bits through a new AMI Reserved parameter of Type “String” and Usage “Out” for the simulator to calculate BER and effective data rate
  - can also return recovered QAM constellations through another new AMI Reserved parameter of Type “String” and Usage “Out” for post-processing such as constellation diagram generation



# DMT AMI Simulation Examples

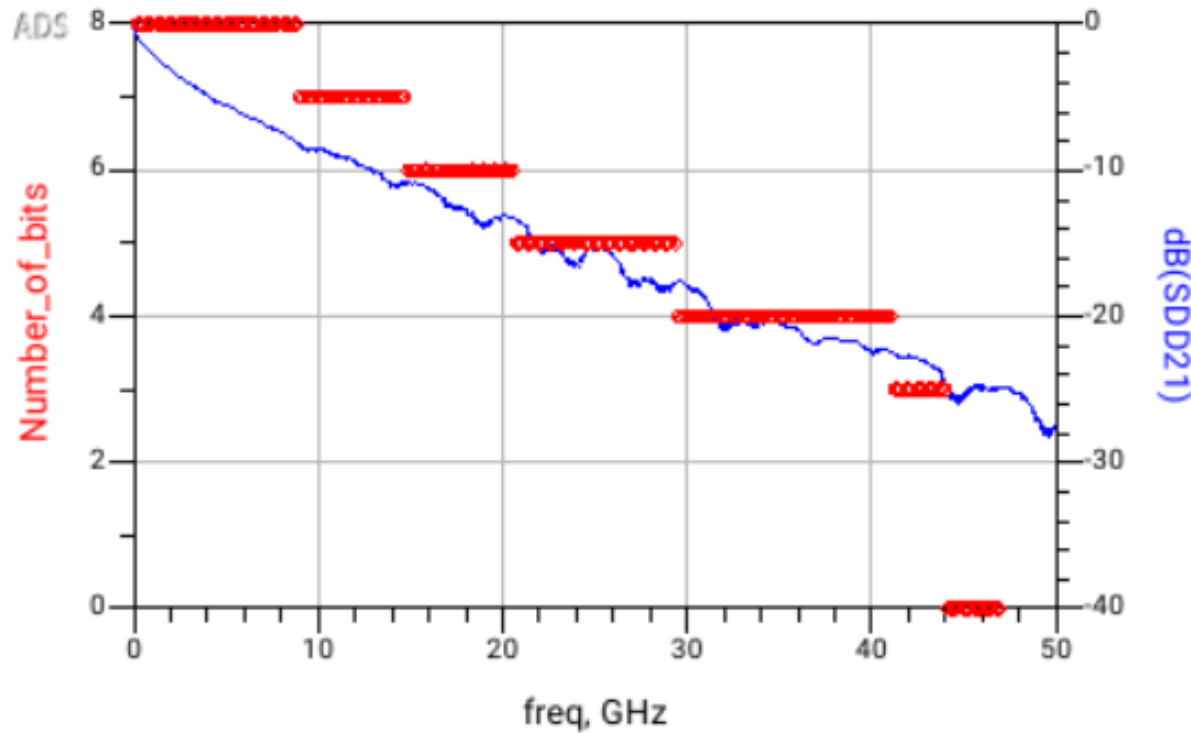
- Target data rate: 200 and 240 Gbps for 100 and 120 GHz sample rates, respectively
- 26 dB differential insertion loss at 50 GHz
- 16 frequency bands with 15 sub-channels per band
- Number of sub-channels:  $16 \times 15 = 240$
- Same QAM order for all 15 sub-channels in each band
- FFT size: 512 (16 unused FFT frequencies including DC)
- CP length: 64 (12.5% of FFT size)
- TX model
  - 9-bit DAC with a full swing of 1 V<sub>dpp</sub>
- RX model
  - 8-bit ADC
  - Applies AWGN (additive white Gaussian noise) at RX input before AGC
  - Noise RMS: 2.5 and 2 mV at 100 and 120 GHz sample rates, respectively



# Results of 100 GHz Sample Rate (1)

- QAM order setting for frequency bands

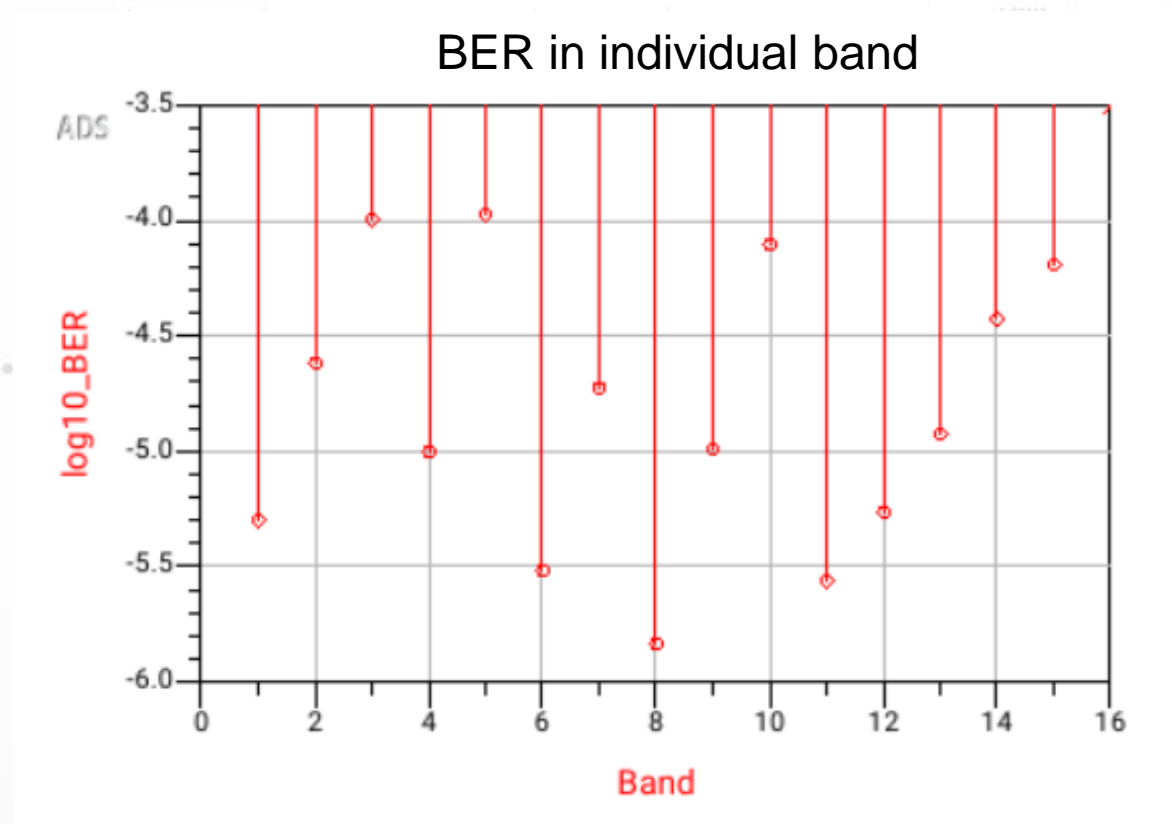
Bands	1-3	4, 5	6, 7	8-10	11-14	15	16
Number of bits per QAM symbol	8	7	6	5	4	3	0



\* Band 16 is not used

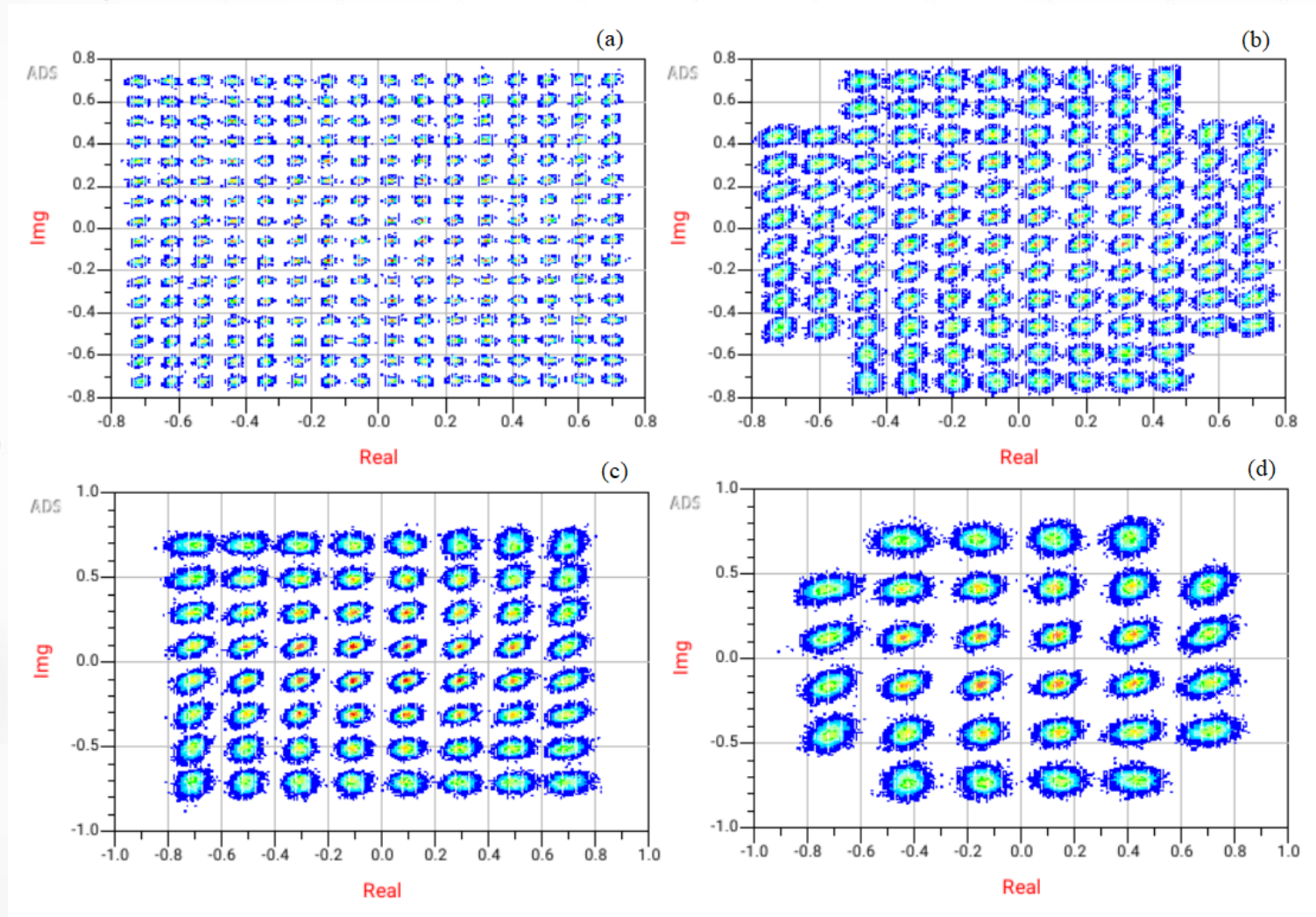
# Results of 100 GHz Sample Rate (2)

- Actual data rate achieved: 200.08 Gbps
- Overall BER:  $3.42 \times 10^{-5}$



# Results of 100 GHz Sample Rate (3)

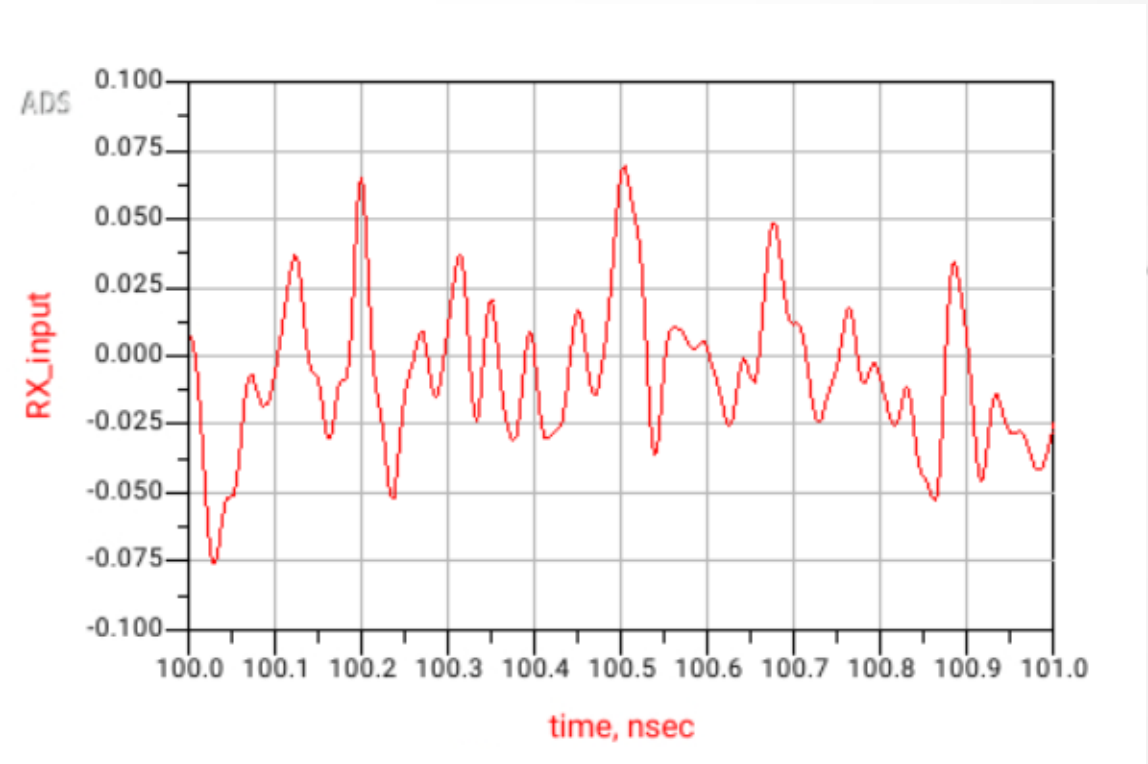
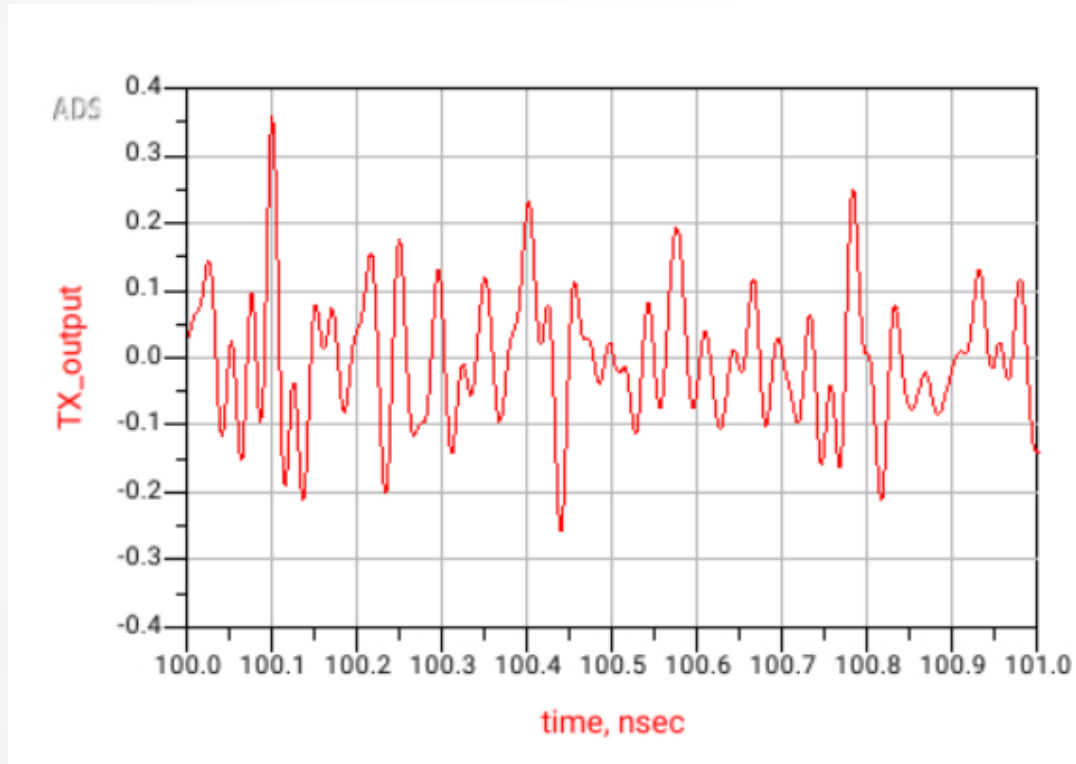
- Constellation diagrams of (a) Band 1, (b) Band 5, (c) Band 7, and (d) Band 9





# Results of 100 GHz Sample Rate (4)

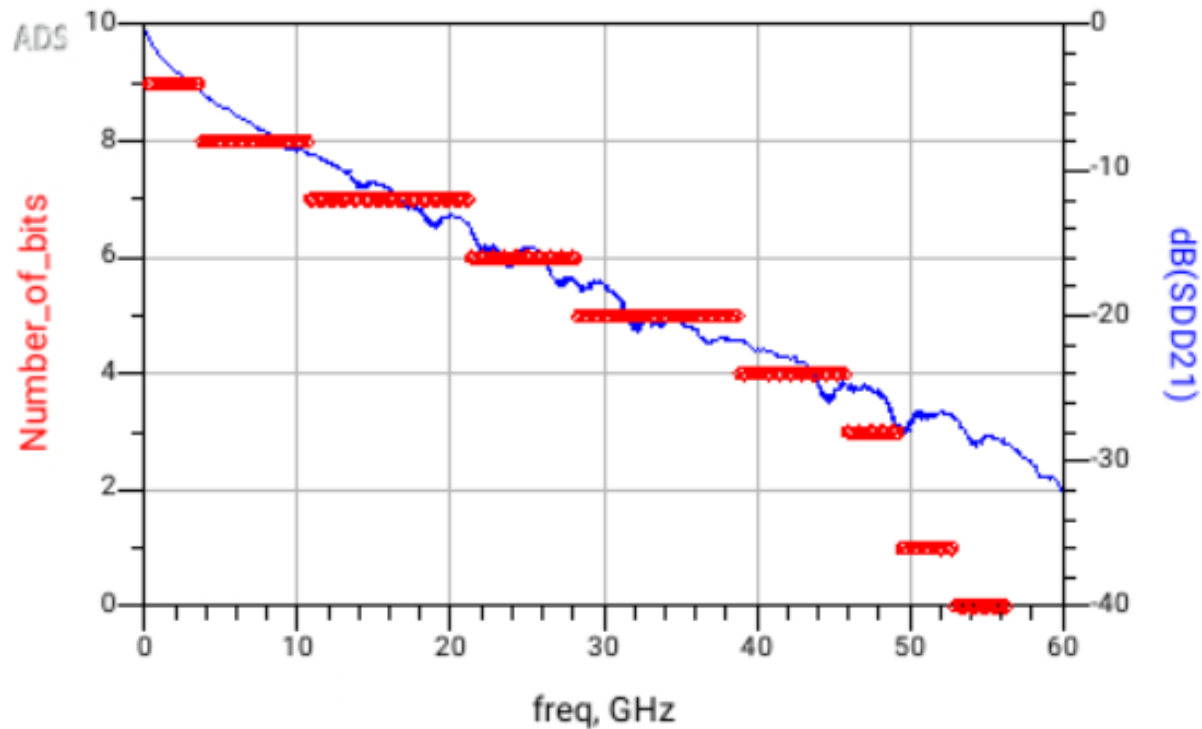
- Waveforms of TX output and RX input



# Results of 120 GHz Sample Rate (1)

- QAM order setting for frequency bands

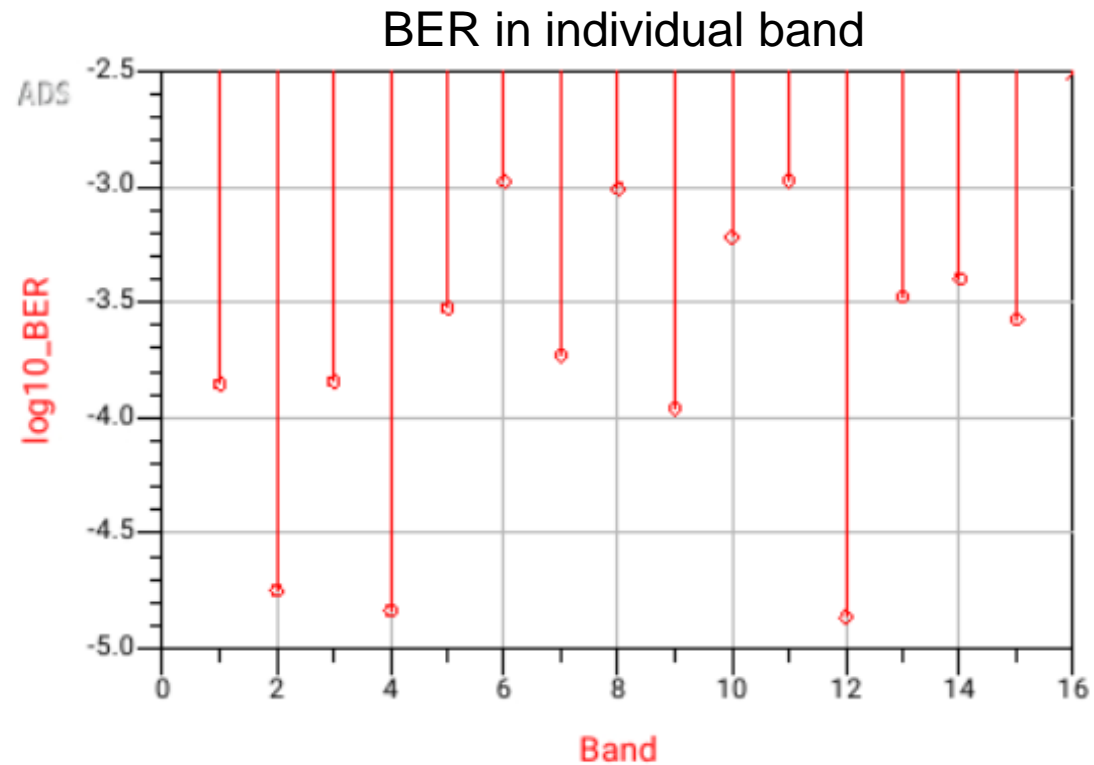
Band	1	2, 3	4-6	7, 8	9-11	12, 13	14	15	16
Number of bits per QAM symbol	9	8	7	6	5	4	3	1	0



\* Band 16 is not used

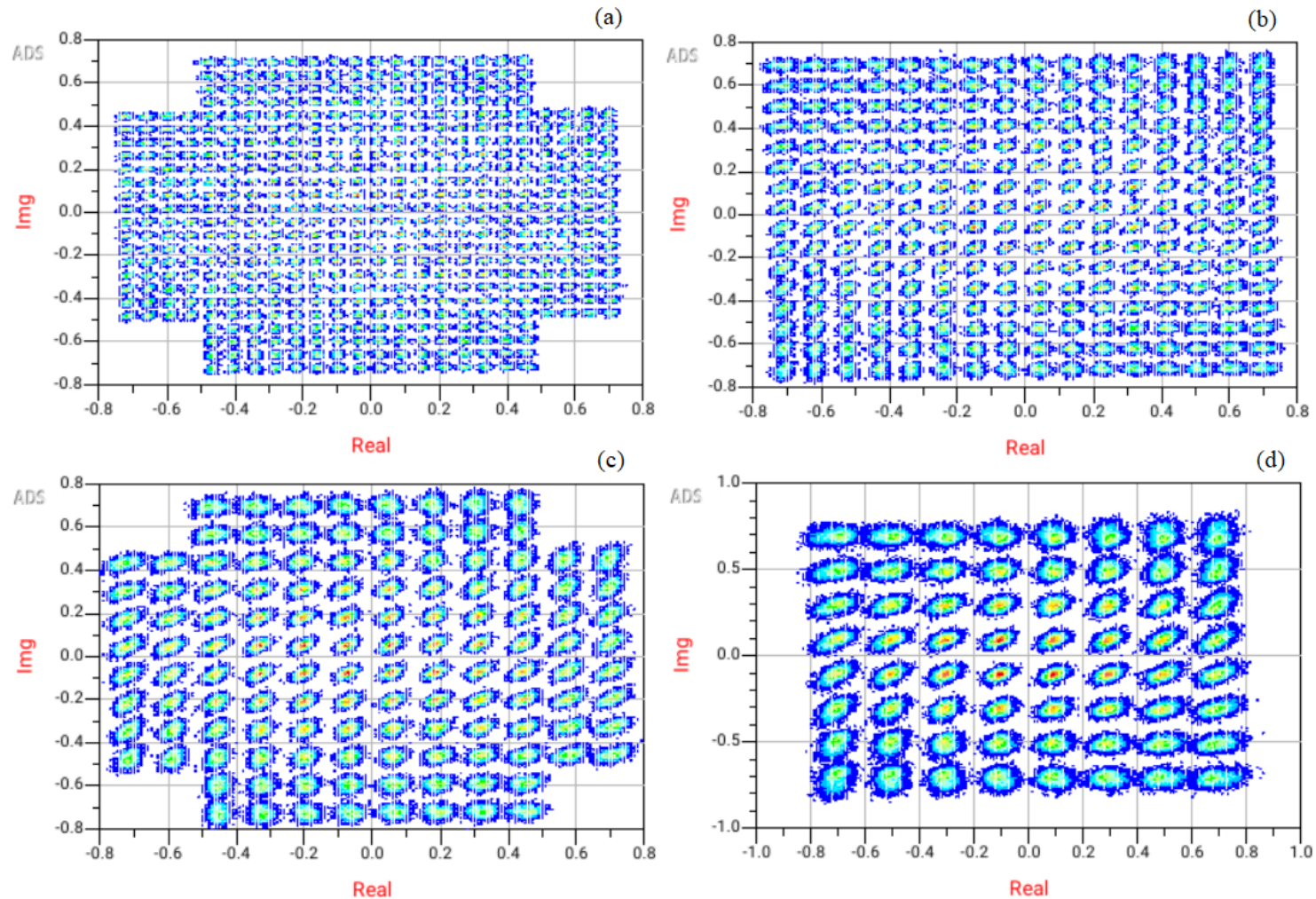
## Results of 120 GHz Sample Rate (2)

- Actual data rate achieved: 242.96 Gbps
- Overall BER:  $3.64 \times 10^{-4}$



# Results of 120 GHz Sample Rate (3)

- Constellation diagrams of (a) Band 1, (b) Band 3, (c) Band 5, and (d) Band 7



# Results of Uniform QAM Modulation (1)

- Uniform QAM order across all 240 sub-channels
- QAM16, QAM32 and QAM64 are studied
- 100 GHz sample rate:

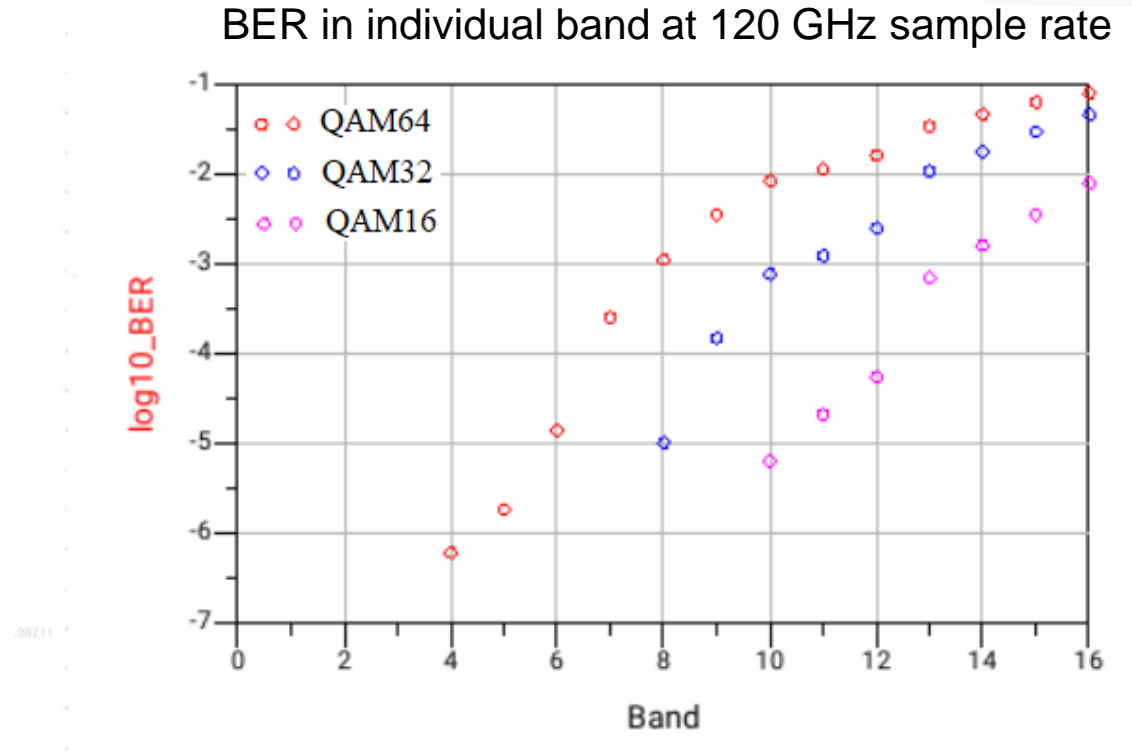
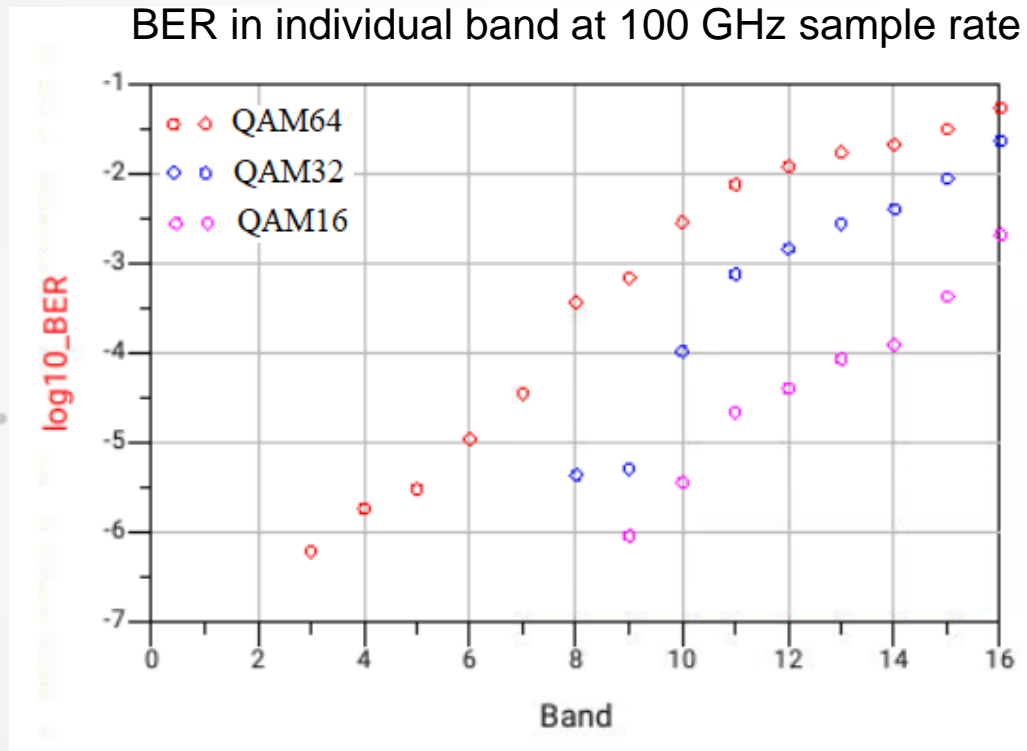
	Uniform QAM16	Uniform QAM32	Uniform QAM64	With bit loading
Overall BER	$1.76 \times 10^{-4}$	$2.61 \times 10^{-3}$	$9.37 \times 10^{-3}$	$3.42 \times 10^{-5}$
Actual data rate	152.44 Gbps	190.56 Gbps	228.67 Gbps	200.08 Gbps

- 120 GHz sample rate:

	Uniform QAM16	Uniform QAM32	Uniform QAM64	With bit loading
Overall BER	$8.72 \times 10^{-4}$	$6.88 \times 10^{-3}$	$1.67 \times 10^{-2}$	$3.64 \times 10^{-4}$
Actual data rate	182.93 Gbps	228.67 Gbps	274.40 Gbps	242.96 Gbps

- At both sample rates, uniform QAM yields worse BER than bit loading does
- The higher the QAM order, the worse the BER

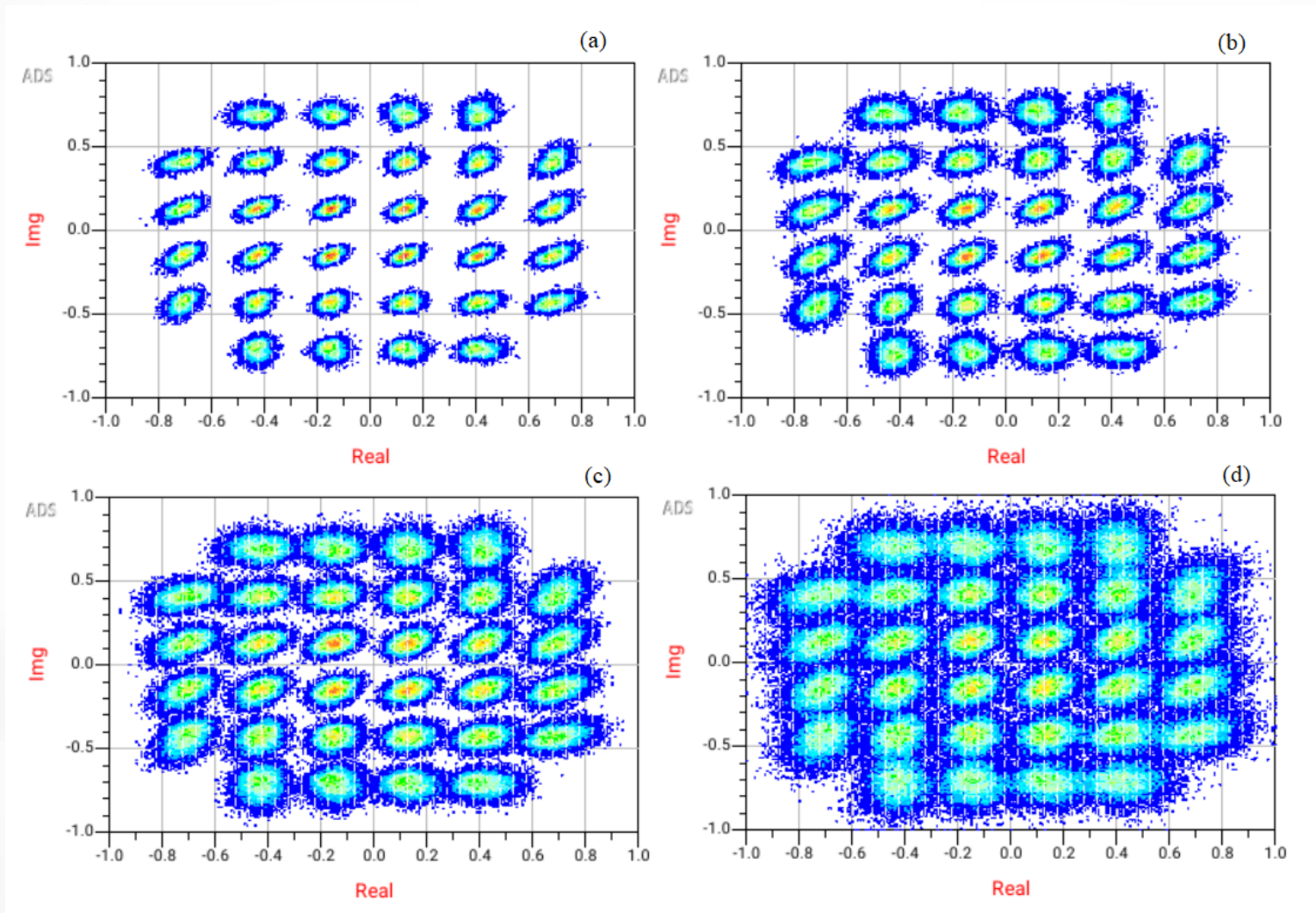
# Results of Uniform QAM Modulation (2)



- The higher the sub-channel frequency, the higher the channel loss, the lower the SNR, the worse the BER in the frequency band

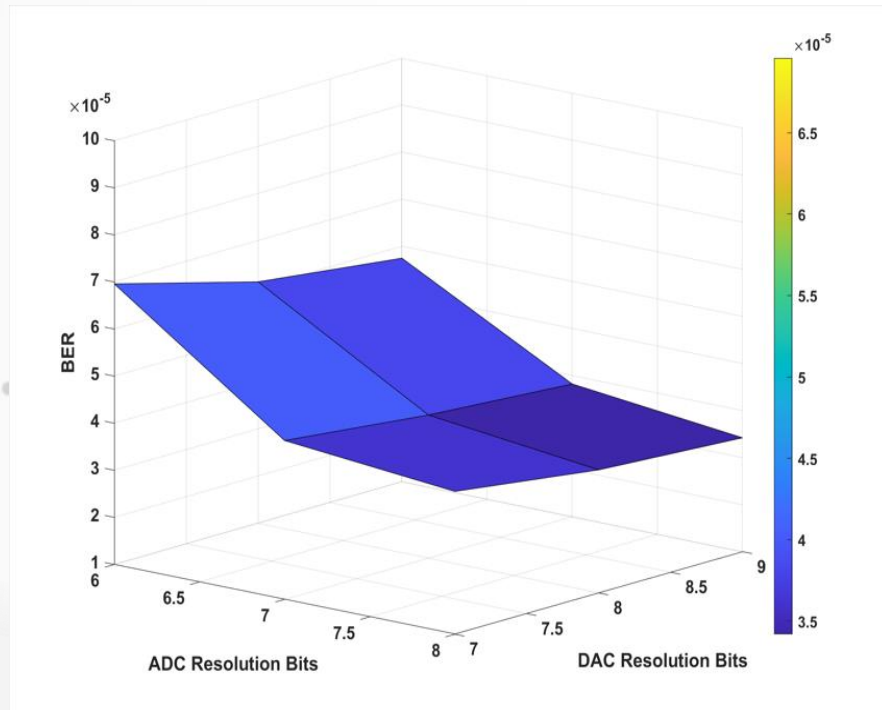
# Results of Uniform QAM Modulation (3)

- Constellations of uniform QAM32 at 120 GHz sample rate. (a) Band 7 (b) Band 9 (c) Band 11 (d) Band 13

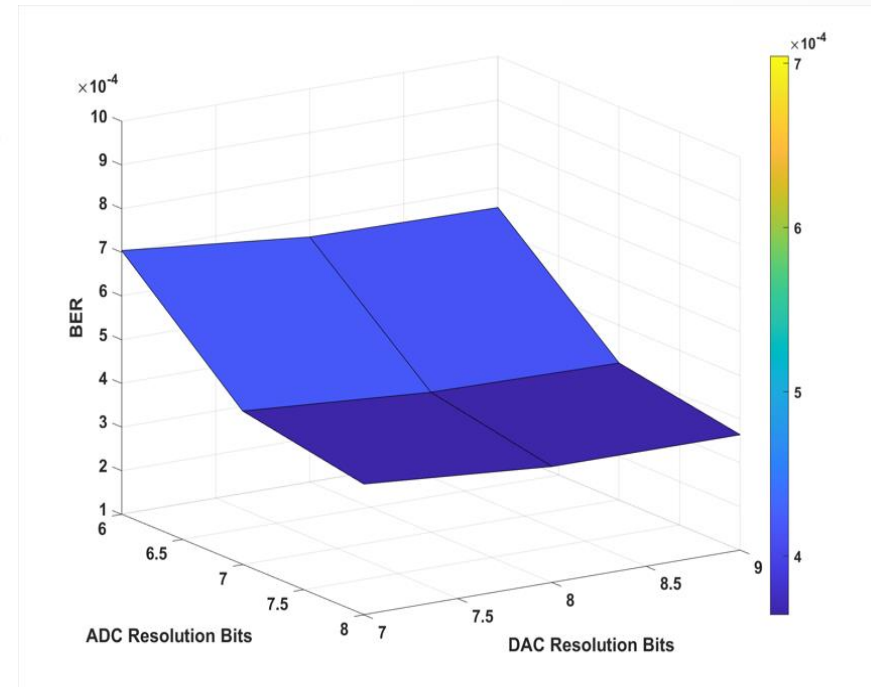


# Impact of DAC and ADC Resolutions (1)

- Sweep TX DAC resolution from 7-bit to 9-bit and RX ADC resolution from 6-bit to 8-bit
- Same bit loading as used previously



BER vs resolutions at 100 GHz sample rate



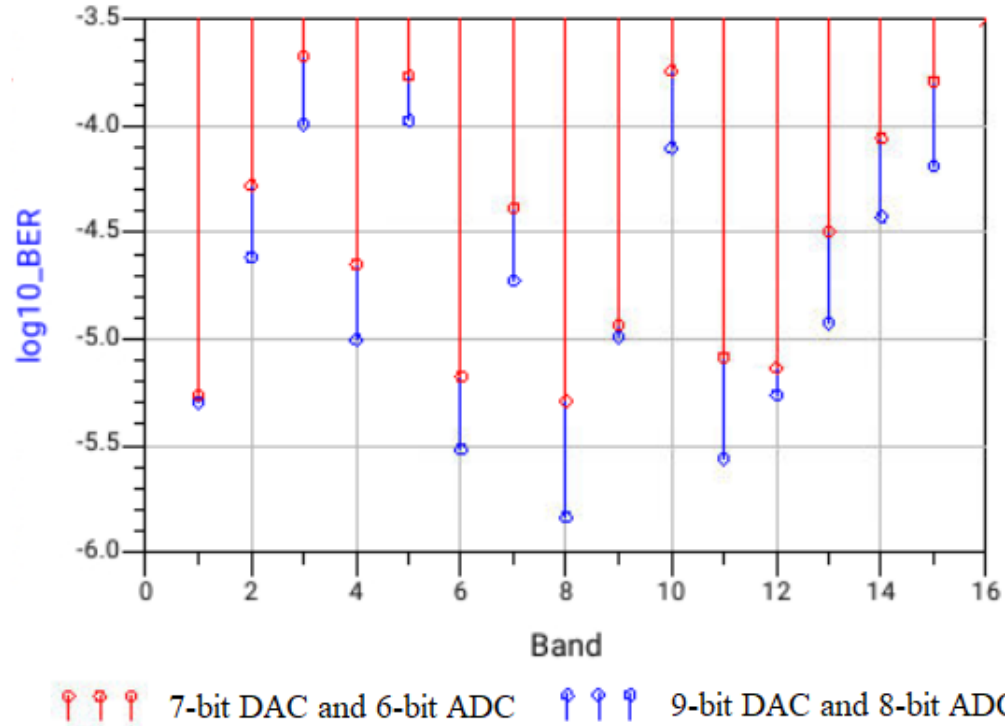
BER vs resolutions at 120 GHz sample rate

- BER improves as DAC and ADC resolutions increase

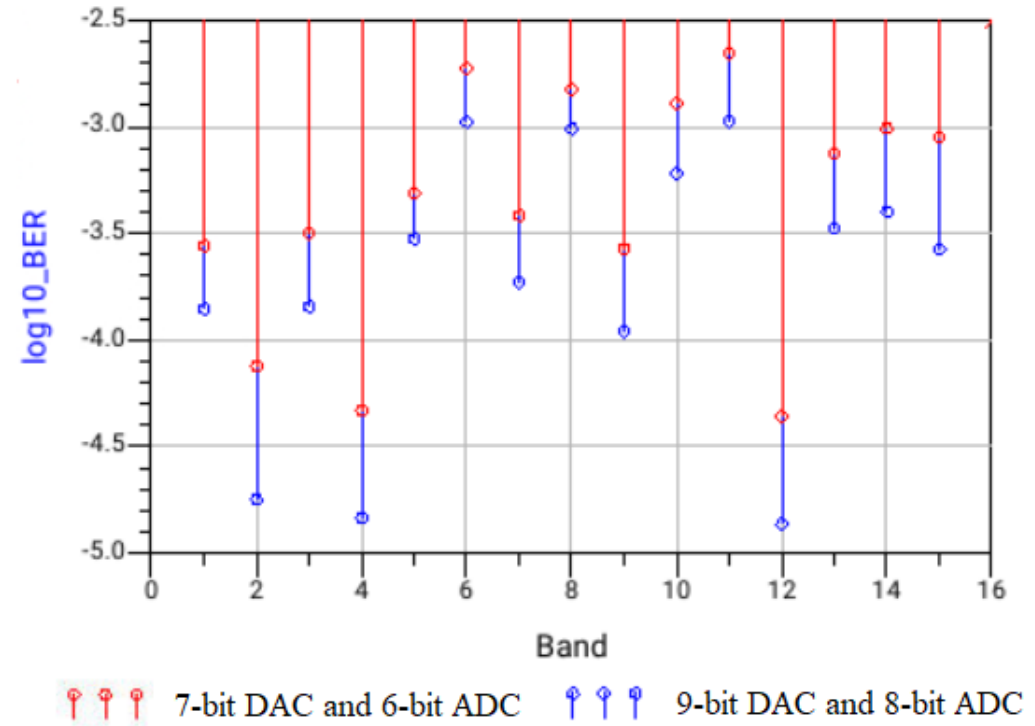


# Impact of DAC and ADC Resolutions (2)

BER in individual band at 100 GHz sample rate



BER in individual band at 120 GHz sample rate

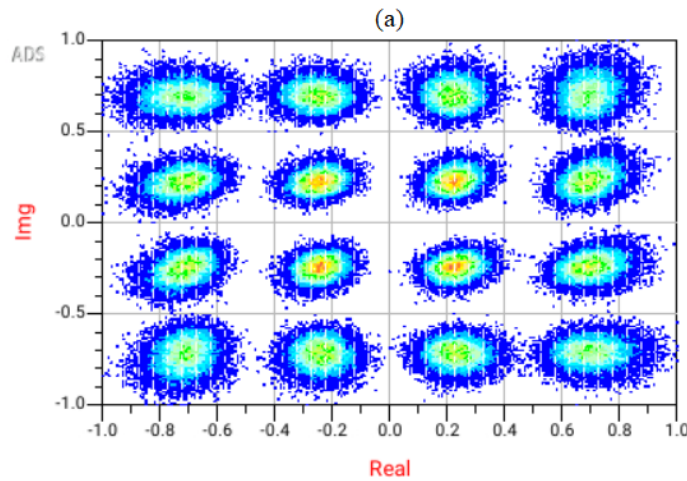


- Lower resolution leads to higher BER in all frequency bands

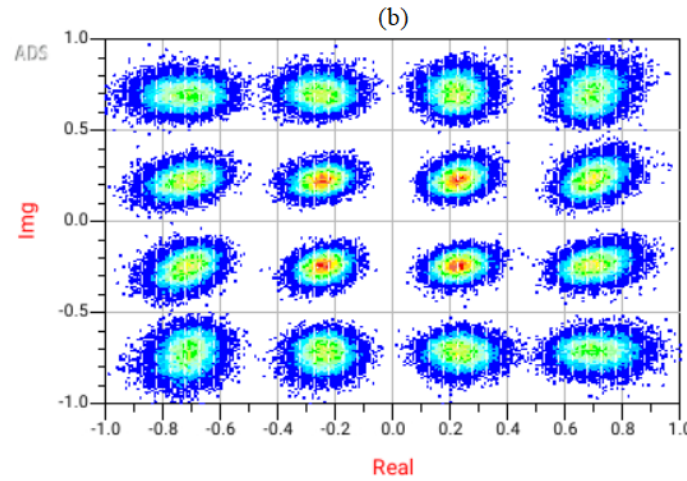
# Impact of DAC and ADC Resolutions (3)

Band 13, 100 GHz  
sample rate

(a) 7-bit DAC and 6-bit ADC

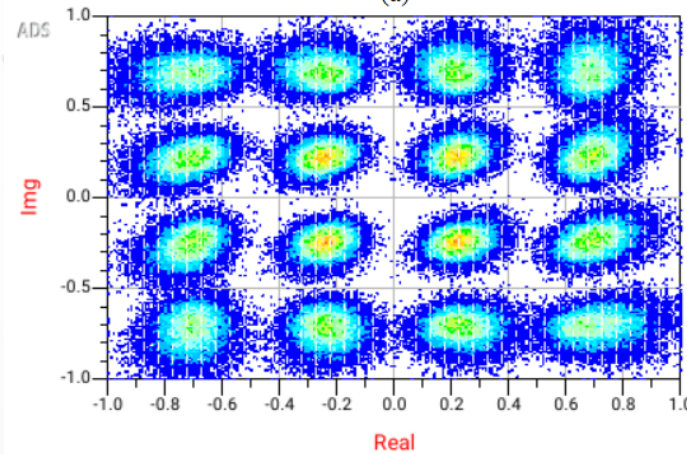


(b) 9-bit DAC and 8-bit ADC

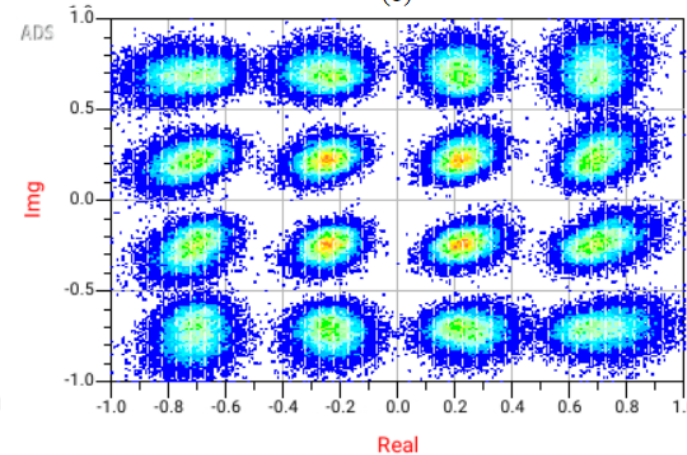


Band 13, 120 GHz  
sample rate

(a)



(b)



- QAM signal noise is higher at lower DAC and ADC resolutions, resulting in worse BER

# Summary

- DMT working principle is briefly discussed in preparation for the AMI modeling
- IBIS-AMI methodology is extended to enable DMT modeling and simulation
  - TX DLL models bit-to-QAM conversion, IFFT, CP insertion and DAC; The TX AMI\_GetWave input is a waveform of binary sequence generated by the simulator, and the output is the waveform of DMT symbols converted from the input bits
  - RX DLL models sample phase detection, ADC, CP removal, FFT, equalization, and QAM-to-bit conversion; RX AMI\_GetWave returns recovered bits and QAM constellations
  - During model initialization, TX AMI\_Init passes all configuration data required by the RX model such as sample rate, number of sub-channels, bit loading, bit-to-QAM mapping, DMT symbol length and CP length to RX AMI\_Init via a proprietary file without the involvement of the simulator
- Examples of DMT with IBIS-AMI modeling are demonstrated
  - Simulated BERs and constellation diagrams are presented and discussed
  - The results show that the system performance is improved by bit loading and higher resolutions of DAC and ADC

**Thank you!**